

**Department of Communications  
Engineering**

Communication Systems

Third Year Class

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Lecture 7

**Types of FM and PM  
Modulations and Generations**

# Types of FM modulation

\* There are two types of FM signals :-

- ① narrow band FM
- ② wide band FM.

Narrowband FM we know

$$S(t)_{FM} = A_c \cos\left[\omega_c t + k_f \int_0^t x(t) dt\right]$$

$$\text{Let } y(t) = \int_0^t x(t) dt$$

$$\therefore S(t)_{FM} = A_c \cos\left[\omega_c t + k_f y(t)\right]$$

Now in phasor form

$$S(t)_{FM} = \text{Re} \left[ A_c e^{j[\omega_c t + k_f y(t)]} \right]$$

OR

$$C(t)_{FM} = A_c e^{j[\omega_c t + k_f y(t)]}$$

\* The condition of narrowband is

$$k_f y(t) \ll 1$$

$$e^{jk_f y(t)} \approx 1 + jk_f y(t)$$

$$C_{FM}(t) = A_c e^{j\omega_c t} \cdot e^{jk_f y(t)}$$

$$= A_c [1 + jk_f y(t)] e^{j\omega_c t}$$

But  $s(t)_{FM} = \text{Real part of } C_{FM}(t)$

$$s(t)_{FM} = A_c \cos \omega_c t - A_c k_f y(t) \sin \omega_c t$$

narrowband PM

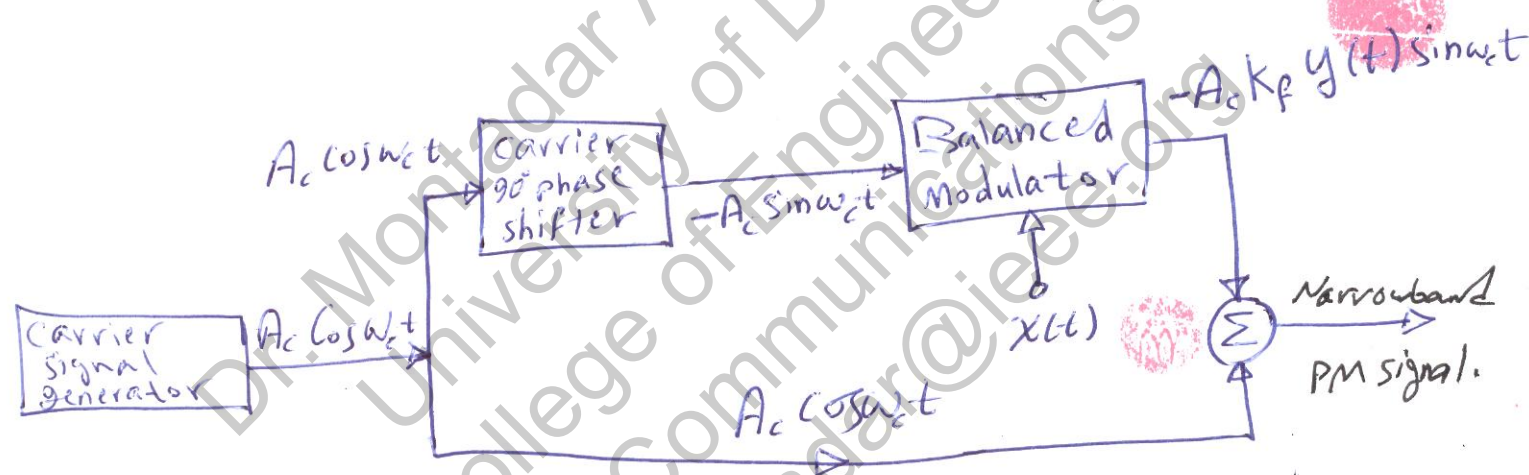
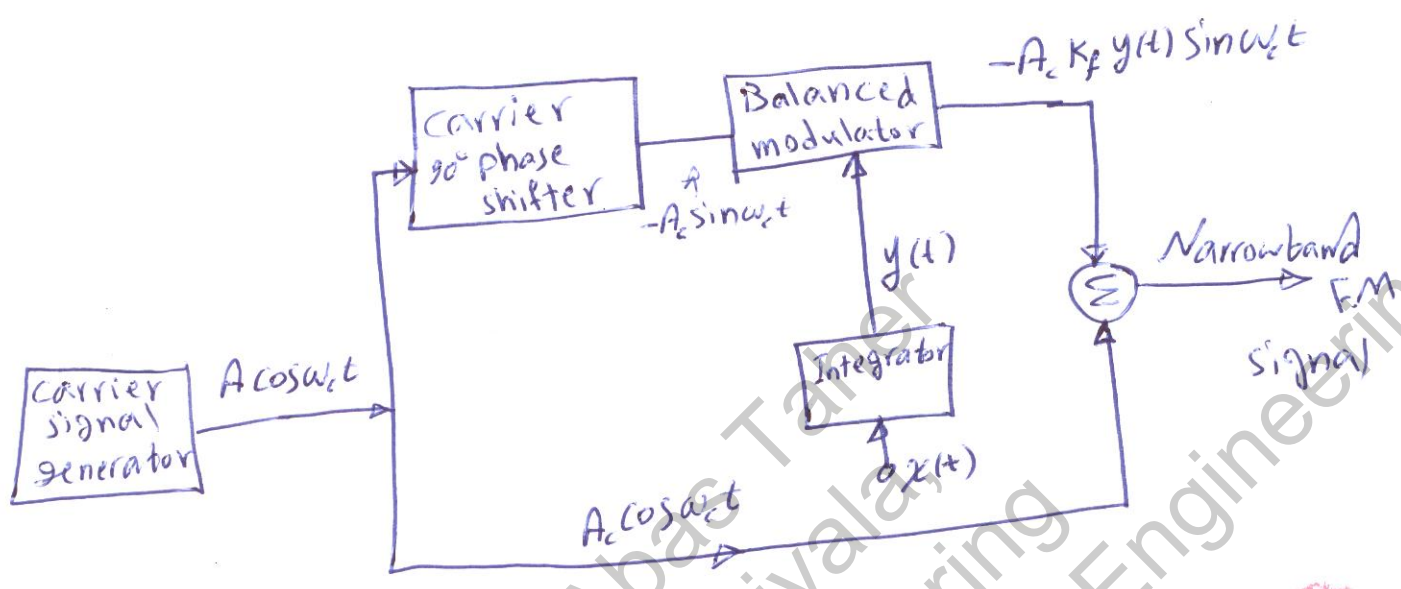
Using the same procedure of the narrowband FM, the PM signal (narrowband) will be

$$s(t)_{PM} = A_c \cos \omega_c t - A k_p x(t) \sin \omega_c t$$

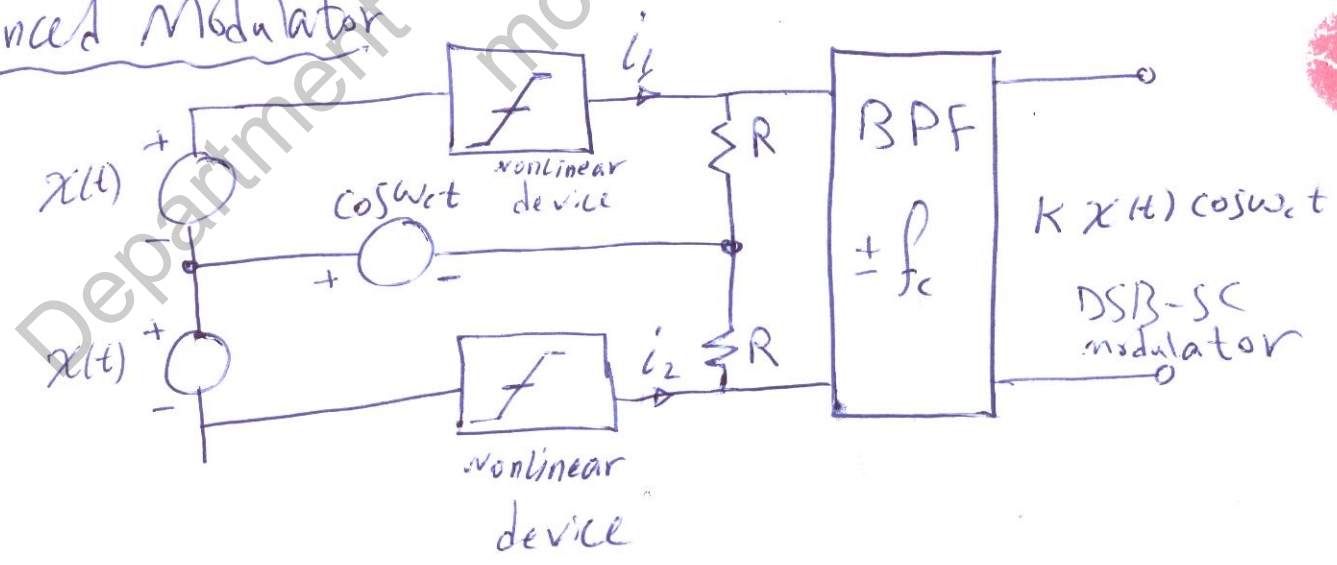
Conclusion for narrowband FM & PM

- It can be seen that the expressions for narrowband FM & PM are similar to the AM signal, thus, the bandwidth is almost similar or same as that of AM.

# Generation of Narrowband FM & PM



## Balanced Modulator



# Single-Tone narrowband FM

$$S(t)_{NFM} = A_c \cos \omega_c t - A_c k_f y(t) \sin \omega_c t$$

$$\text{and } y(t) = \int x(t) dt$$

$$\text{let } x(t) = V_m \cos \omega_m t \quad (\text{the message signal})$$

$$y(t) = \int V_m \cos \omega_m t dt = \frac{V_m}{\omega_m} \sin \omega_m t$$

$$S(t)_{NFM} = A_c \cos \omega_c t - A_c k_f \frac{V_m}{\omega_m} \sin \omega_m t \sin \omega_c t$$

$$\text{but } m_f = \frac{k_f V_m}{\omega_m}$$

$$S(t)_{NFM} = A_c \cos \omega_c t - A_c m_f \sin \omega_m t \sin \omega_c t$$

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## Wideband FM

\* if  $m_f$  is large  $\rightarrow$  large number of sidebands produced

\* To understand these complicated analysis, a message of single tone sinusoid will be used.

$$s(t) = A \cos(\omega_c t + m_f \sin \omega_m t)$$

The phasor form of  $s(t)$  is

$$C_{FM}(t) = A e^{j\omega_c t} e^{j m_f \sin \omega_m t}$$

periodic with period  $\frac{1}{f_m}$

\* thus  $e^{j m_f \sin \omega_m t}$  is periodic with period  $\frac{1}{f_m}$

\*  $e^{j m_f \sin \omega_m t}$  can be expanded using Fourier series in

$$e^{j m_f \sin \omega_m t} = \sum_{n=-\infty}^{\infty} C_n e^{j n \omega_m t} \quad \text{for } -\frac{1}{2f_m} \leq t \leq \frac{1}{2f_m}$$

$$C_n = f_m \int_{-\frac{\pi}{\omega_m}}^{\frac{\pi}{\omega_m}} e^{j(m_f \sin \omega_m t)} e^{-j n \omega_m t} dt$$

$$= \frac{1}{2\pi} \int_{-\pi}^{\pi} e^{j(m_f \sin x - n x)} dx \quad \text{where } x = \omega_m t$$

$$C_n = J_n(m_f)$$

where  $J_n(m_f)$  is the Bessel function of  $n^{\text{th}}$  order of the first kind.

$$e^{jm_f \sin \omega_m t} = \sum_{n=-\infty}^{\infty} J_n(m_f) e^{jn \omega_m t}$$

Thus

$$C_{FM}(t) = A e^{j\omega_c t} \sum_{n=-\infty}^{\infty} J_n(m_f) e^{jn \omega_m t}$$

$$C_{FM}(t) = A \sum_{n=-\infty}^{\infty} J_n(m_f) e^{j(\omega_c + n\omega_m)t}$$

from which

$$S(t)_{FM} = A \sum_{n=-\infty}^{\infty} J_n(m_f) \cos(\omega_c t + n\omega_m t)$$

\* Properties of Bessel function

$$\textcircled{1} \begin{aligned} J_n(m_f) &= J_{-n}(m_f) & n \text{ is even} \\ J_n(m_f) &= -J_{-n}(m_f) & n \text{ is odd} \end{aligned}$$

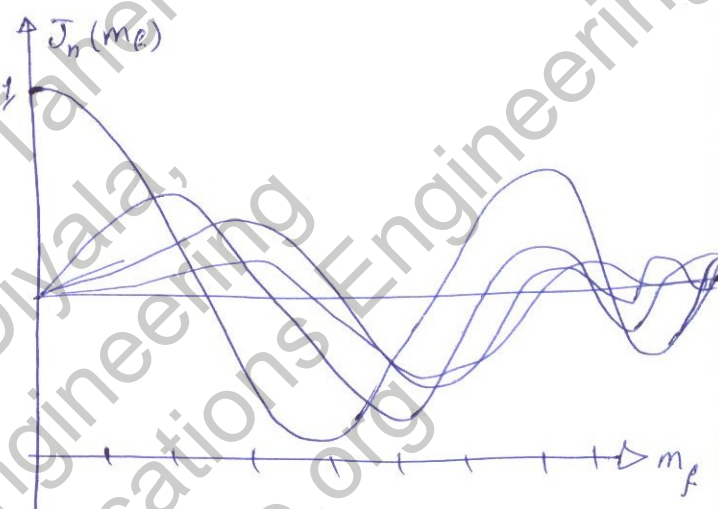
$$\textcircled{2} \text{ For small values of } m_f :- J_0(m_f) \approx 1, J_1(m_f) \approx \frac{m_f}{2}, J_n(m_f) \approx 0 \quad n > 1$$

$$\textcircled{3} \sum_{n=-\infty}^{\infty} J_n^2(m_f) = 1$$

using the first property:

$$\begin{aligned}
S_{FM}(t) = & A J_0(m_f) \cos \omega_c t + A J_1(m_f) [\cos(\omega_c + \omega_m)t - \cos(\omega_c - \omega_m)t] \\
& + A J_2(m_f) [\cos(\omega_c + 2\omega_m)t + \cos(\omega_c - 2\omega_m)t] \\
& + A J_3(m_f) [\cos(\omega_c + 3\omega_m)t - \cos(\omega_c - 3\omega_m)t] \\
& + \dots
\end{aligned}$$

\* we conclude from the last expression of  $S_{FM}(t)$



① infinite Bandwidth (theoretically)

because of the infinite number of sidebands

② For small  $m_f$  (less than 0.6)

- thus there is only the carrier term and one pair of sidebands.
- this case equivalent to narrowband FM.

$$S_{FM}(t) = A J_0(m_f) \cos \omega_c t + A J_1(m_f) [\cos(\omega_c + \omega_m)t - \cos(\omega_c - \omega_m)t]$$

③ If  $m_f > 1$  to be  $\approx 2.4$  or  $5.2$ ,  $J_0(m_f) = 0$ ,

∴ carrier power = 0, all power carried by the ~~modulated~~ sidebands

~~signal~~ Hence efficiency = 100%

∴  $33 \text{ AM} \ll \sum_{FM} \ll \text{DSB-SC } 100\%$



Thus: According to CCIR (consultative committee for International Radio), there are some regulations:-

- ① Maximum modulation Frequency = 15 kHz, (message)
- ② Maximum frequency deviation  $\Delta f = 75$  kHz,
- ③ Frequency stability of the carrier is  $\pm 2$  kHz,
- ④ Allowable bandwidth per channel = 200 kHz.

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